

**CORDLESS TELEPHONE-TO-SOUND CARD INTERFACE ADAPTER
HAVING A HYBRID TRANSFORMER CIRCUIT**

5 **CROSS-REFERENCE TO RELATED APPLICATION**

This application claims priority to Mexican patent application entitled "Device Allowing Connection of Wireless Telephone to Computer for Internet Telephony" having Serial No. PA/a/2003/005992 and filing date of July 2nd 2003.

10 **BACKGROUND**

Field of the Invention

The present invention relates to a cordless telephone-to-sound card interface adapter having a hybrid transformer circuit, which provides mobility to an end user during voice communications over the Internet.

15 **Description of the Related Art**

Historically, a determining factor in the low usage of Internet telephony usage was attributed to poor voice quality. Nevertheless, in the last few years major progress has been made in this area to address the voice quality issues. These improvements have motivated
20 Internet users, especially people or companies with limited budgets for international long-distance calls, to take advantage of the service. The service advantageously provides savings of up to 70% over the traditional calling system, in addition to the relatively low rates offered by Internet providers and traditional local-call telephony.

The present application concerns an interface adapter which connects between a
25 cordless telephone system and a computer for facilitating voice communications over the Internet using Voice over IP (VoIP) communications. Currently, there are at least two such interface adapter devices on the U.S. market – namely, the Phonebridge device and the ChatBridge device. The ChatBridge device utilizes a Universal Serial Bus (USB) interface, a digital signal processor (DSP), and sophisticated software to emulate an application on the
30 computer for canceling echo so that a sound card is not needed for voice communication. With the Phonebridge device, the computer must be equipped with a sound card to establish voice communication but sophisticated electronics are also required for its operation.

What is needed is a low-cost but effective interface adapter which connects between a cordless telephone and a computer for facilitating voice communications over the Internet using VoIP communications.

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SUMMARY

What I have developed is a cordless telephone-to-sound card interface adapter that will allow Internet users to establish PC-to-telephone and PC-to-PC voice conversations over the Internet by means of instant messaging using their PC and any standard cordless telephone, freeing them from sitting idly in front of their PC and offering mobility while they
10 talk anywhere in their homes. The interface adapter utilizes passive components for its operation, which makes it more affordable and leads to fewer risks of electrical discharges. Preferably, the interface adapter includes a hybrid transformer circuit for matching the impedance between a PC sound card (600 Ω impedance of the speaker line and 10K Ω impedance of the microphone line) and the cordless telephone system (450 Ω nominal
15 impedance), thereby canceling out voice echo during telephone conversations over the Internet. Advantageously, the interface adapter provides a low-cost yet effective means to provide mobility for an end user during voice communications over the Internet.

BRIEF DESCRIPTION OF THE DRAWINGS

20 Embodiments of present invention will now be described by way of example with reference to attached figures, wherein:

FIG. 1 is a block diagram illustration of a voice communication system which utilizes a cordless telephone-to-sound card interface adapter of the present application;

FIG. 2 is a schematic diagram of a hybrid transformer circuit of the interface adapter
25 of FIG. 1 in a first embodiment;

FIG. 3 is a schematic diagram of the hybrid transformer circuit of FIG. 2, showing relevant equivalent circuit information when coupled to a cordless telephone system and a sound card of a computer;

FIG. 4 is a schematic diagram of the hybrid transformer circuit for the interface
30 adapter of FIG. 1 in a second embodiment;

FIG. 5 is the schematic diagram of the hybrid transformer circuit of FIG. 4 which further includes a voltage bias interface; and

FIG. 6 is a block diagram illustration of the two hybrid transformers of the hybrid transformer circuit, which are physically oriented 90° out-of-phase in the housing so as to
 5 reduce electromagnetic interference with one another.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

As described herein, a cordless telephone-to-sound card interface adapter of the present application is configured to operate with any PC equipped with a sound card and any
 10 standard telephone, which is preferably a cordless telephone system to provide the user with mobility. In one specific implementation, the interface adapter includes a hybrid transformer circuit having two hybrid transformers with multiple 600 Ω coils and an impedance matching circuit having a precision resistor and an optional capacitor. The primary function of the hybrid transformer circuit, which operates through electromagnetic coupling like other
 15 transformers, is to effect the conversion from two wires (the cordless telephone) to four wires (the computer sound card) with suitable impedance matching, thereby separating the transmitted voice from the received voice without producing voice echo toward the origin. The interface adapter includes a speaker plug for a receive signal (Rx) and a microphone plug for a transmit signal (Tx) which are connected to the computer's sound card. A
 20 telephone interface of the interface adapter is connected to the cordless telephone system to facilitate voice conversation over the Internet with end-user mobility. The interface adapter may further include a Universal Serial Bus (USB) plug for biasing circuits of the cordless telephone system through the interface adapter. Advantageously, the interface adapter of the present application is a simple low-cost means for providing mobility to an end user for voice
 25 communications over the Internet. With convenient cordless telephone mobility, this adapter facilitates significant savings in long distance telephone services.

FIG. 1 is an illustration of a voice communication system 100 which utilizes a cordless telephone-to-sound card interface adapter 102 of the present application. In general, interface adapter 102 helps facilitate voice communications between a first end user of a
 30 cordless telephone system 108 and a second end user of a telephone 136/148 or a computer 142 using instant messaging communications over the Internet.

Interface adapter 102 includes a hybrid transformer circuit 116 as well as a telephone jack 118, a speaker plug 122, a microphone plug 124, and a Universal Serial Bus (USB) plug 127 which are coupled to hybrid transformer circuit 116. Hybrid transformer circuit 116, which will be described later in detail in relation to FIGs. 2-4, is carried and contained
5 entirely within a housing unit 120 (e.g. a molded hard plastic housing unit). Telephone jack 118 is carried on a side edge of housing unit 120 and is coupled to a conventional cordless telephone system 108 through a telephone cord. Preferably, telephone jack 118 is of the RJ-11 (female) type. Cordless telephone system 108 includes a cordless base unit 110 and at least one cordless handset unit 112. As is conventional, cordless base unit 110 and cordless
10 handset unit 112 communicate via radio frequency (RF) signals 114 to provide mobility to an end user of cordless handset unit 112.

The other end of interface adapter 102 is configured to connect to a standard sound card 106 of a computer 104, such as a Personal Computer (PC). In particular, speaker plug 122 and microphone plug 124 of interface adapter 102 are configured to interface with sound
15 card 106. A speaker plug cable 126 extends from housing unit 120 and terminates with speaker plug 122. Similarly, a microphone plug cable 128 extends from housing unit 120 and terminates with microphone plug 124. Preferably, speaker plug 122 and microphone plug 124 are both male connectors of the 1/8" miniplug type. Sound card 106 has suitable female connectors of the 1/8" minijack type to receive the 1/8" miniplugs from interface
20 adapter 102. In addition, USB plug 127 is configured to interface with a USB jack on the PC for receiving a bias voltage to bias circuits of cordless telephone system 108 through interface adapter 102. Preferably, speaker plug cable 126, microphone plug cable 128, and USB plug cable 125 are actually bound together as a single cable which extends from housing unit 102 between about 24 – 40 inches (i.e. about 61 – 101 cm).

25 Computer 104 connects to the Internet 132 through a local Internet Service Provider (ISP) 133. Sound card 106 of computer 104 converts analog voice signals from interface adapter 102 (which originated from cordless telephone system 108) into digital data for communication over the Internet 132. Conversely, sound card 106 converts digital voice data from the Internet 132 (which originated from telephone 136 or computer 142) into
30 analog voice signals to interface adapter 102. A software application (not shown) which resides on computer 104 is utilized to help facilitate an instant voice messaging service, such

as that service provided by MSNTM Messenger, America On-Line (AOL)TM Instant MessengerTM, or YahooTM Messenger, or SkypeTM, which is based on Voice over IP (VoIP) technology.

5 Elements which exist outside of the home or office of the end user of cordless telephone system 108, or outside of the end user's purview 130, include the Internet 132, ISPs 133 and 146, telephone network 134, Internet-to-phone network server/gateway 138, as well as other end user's telephones 136 and 148, interface adapter 140, and computer 142. The other end user in the voice conversation may employ one of three alternatives: (1) telephone 136 via the standard telephone network 134; (2) computer 142 via the Internet 132
10 using a standard speaker/microphone attachment (attachment not shown) to computer 142; or (3) telephone 148 via the Internet 132 which is coupled to computer 142 via interface adapter 140.

If computer 142 with the speaker/mic attachment is utilized, the speaker/mic attachment is coupled to a sound card 144 of computer 142 which is coupled to the Internet
15 132 through local ISP 146. In this scenario, the voice communication does not require the use of telephone network 134 at all, except for any telephone connections established between computers 104/142 and their corresponding ISPs 133/146. The likelihood of undesirable echo generation in this scenario is very high if no echo cancellation is utilized by computer 104, since any voice signals from the speaker line will be fed to the microphone
20 line and back to the origin. Suitable standard headphone sets are recommended for use so that acoustical echo will be eliminated.

On the other hand, telephone 148 and interface adapter 140 may be utilized with computer 142. Interface adapter 140 is similar in make and functionality as interface adapter 102 and, preferably, telephone 148 is a cordless telephone system (i.e. like cordless telephone
25 system 108). In this scenario, the voice communication also does not require the use of telephone network 134 at all, except for any telephone connections established between computers 104/142 and their corresponding ISPs 133/146.

Using the last alternative, the end user of cordless telephone system 108 may have a subscription with an Internet-to-telephone network service provider, such as Net2Phone of
30 Newark, New Jersey, U.S.A.; Callserve of London in the U.K.; or iConnectHere of DeltaThree, Inc. Such a service utilizes an Internet-to-telephone network server or gateway

138 and/or other devices for converting VoIP data packets from the Internet 132 into conventional telephone signals for use over telephone network 134, and vice versa. With such a subscription, the end user of cordless telephone system 108 may communicate over the Internet 132 and telephone network 134 with any other telephone user including the end
 5 user of telephone 136.

FIG. 2 is a schematic diagram of hybrid transformer circuit 116 of FIG. 1. Hybrid transformer circuit 116 of FIG. 2 includes a first hybrid transformer 202, a second hybrid transformer 204, and an impedance matching circuit 206. Hybrid transformers are used to convert a bi-directional two-wire signal into one-way four-wire signals. In the embodiment
 10 of FIG. 2, each hybrid transformer 202 and 204 is a $600\ \Omega - 600\ \Omega$ single-hybrid transformer which connects a two-wire port (telephone set) to a four-wire transmit and receive port (PC sound card). Preferably, hybrid transformers 202 and 204 are single hybrid transformers of model SPT-180 available from Prem Magnetics, Inc., located in McHenry, Illinois, U.S.A. Mouser Electronics located in Mansfield, Texas, U.S.A. may be another reliable source for
 15 hybrid transformers (e.g. Model TL028). However, any suitable hybrid transformer from any suitable manufacturer may be utilized. FIG. 3 shows the same hybrid transformer circuit 116 coupled in its environment which is illustrated by equivalent circuit information (i.e. equivalent circuit voltages and impedances).

In FIGs. 2-3, hybrid transformer 202 has a pair of mutually-coupled coils which
 20 include mutually-coupled coils B and D and mutually-coupled coils A and C formed along the same core. Similarly, hybrid transformer 204 has a pair of mutually-coupled coils which include mutually-coupled coils F and H and mutually-coupled coils E and G formed along the same core. A terminal 1 of coil B and a terminal 6 of coil A of hybrid transformer 202 are coupled to speaker plug 122 (FIG. 1), whereas a terminal 1 of coil G and a terminal 6 of
 25 coil H of hybrid transformer 204 are coupled to microphone plug 124 (FIG. 1). A terminal 7 of coil D and a terminal 12 of coil F are coupled to two-line telephone jack 118 (FIG. 1). A terminal 8 of coil D of hybrid transformer 202 is coupled to a terminal 11 of coil F of hybrid transformer 204. A terminal 12 of coil C of hybrid transformer 202 is coupled to a terminal 8 of coil E of hybrid transformer 204, which makes coils C and coils E reverse-coupled. A
 30 terminal 11 of coil C of hybrid transformer 202 and a terminal 7 of coil E of hybrid transformer 204 are coupled to impedance matching circuit 206 which has an impedance Z_B .

In this embodiment, impedance matching circuit 206 includes a resistor 210 and an optional capacitor 208 coupled in a parallel configuration as shown.

Assuming a change of current in coil A, a change will be produced in the magnetic field of the same coil which, in turn, produces a change of voltage in coil C on the same transformer core. When coil C is connected in a circuit, changes in voltage in the circuit will generate a change in current flow. Therefore, the pattern of change of current in the primary coil A is reproduced in the transformer's secondary coil C without both circuits having to be physically connected. Note that the ratio of incoming voltage to outgoing voltage (and of incoming current to outgoing current) depends on the ratio of the number of turns of the primary and secondary windings. This makes it possible to equalize the impedance between the coupled circuits (i.e. the sound card and the cordless telephone) as $N^2 \times Z_p$, where N is the ratio of secondary-to-primary coil turns of the transformer and Z_p is the impedance of the primary coil. Impedance matching is important to reduce the reflection of transmitted signals and corresponding loss of power. Preferably, the ratio N for both transformers 202 and 204 is between 1:1.60 (+/- 2%) using the SPT-180 type transformers.

A voice signal is received from the sound card (i.e. speaker line) that is applied to terminals 1 and 6 of transformer A 202, next to be coupled inductively to its coils C and D owing to the flow of current through coils A and B. The voltage induced in coil D causes current to flow through the two-wire circuit connected to terminal 7 of hybrid transformer 202 and terminal 12 of hybrid transformer 204, to be transmitted further on. This same current also flows through coil F, which causes an induced voltage in coil H. Likewise, the voltage of the voice signal induced in coil C causes current to flow through impedance matching circuit 206 and coil E. The impedance Z_B of impedance matching circuit 206 is tuned with precision to equalize an impedance Z_L from the telephone set line, thereby obtaining zero return of the transmitted signal toward its origin. That is, the impedance Z_B of impedance matching circuit 206 is set during the design phase to match the line impedance Z_L which may vary slightly from device to device. Given that Z_B is equal to Z_L , the number of turns in coils C and D are equal in number, and the number of turns in coils E and F are also equal in number, the same flow of current exists in the circuits of coils C and E and D and F, thereby yielding the same voltage induced in coils G and H (each of which also have the same number of turns).

In general, impedance matching circuit 206 may have an impedance Z_B between about 350 and 500 Ω . Preferably, the impedance Z_B may be set to between about 457 and 467 Ω . Most preferably, a fixed high-precision resistor (e.g. preferably a 1% tolerance or less) having a resistance of about 467 Ω is utilized. Capacitor 208, which is entirely optional,
 5 may have any suitable capacitance (e.g. about 0.2 μF) to filter out any undesirable high-frequency components.

Coils C and E have their connections inverted from the connections of coils D and F so that any voltage induced in coil G is opposite in phase to the voltage induced in coil H, opposing it and resulting in the cancellation of both. The signal appearing at terminals 1 and
 10 6 of hybrid transformer 202 (i.e. speaker line) is transmitted to terminals 7 and 12 without appearing at terminals 1 and 6 of hybrid transformer 204 (i.e. microphone line). The level of the signal transmitted at terminals 7 (hybrid transformer 202) and 12 (hybrid transformer 204) (i.e. the telephone set line) is half the level of the incoming signal at terminals 1 and 6 of hybrid transformer 202 (i.e. speaker line), as the other half of the signal is dissipated in
 15 impedance matching circuit 206. When a voice signal is received from the telephone set at terminals 7 and 12 of hybrid transformers 202 and 204, its magnitude at both the sound card speaker line and the sound card microphone line will be only half of the original signal. The reason is that coil D of hybrid transformer 202 and coil F of hybrid transformer 204 are connected in series with the telephone set as shown. The level of the voice signal injected at
 20 the sound card microphone input (terminals 1 and 6 of hybrid transformer 204) may be increased at the other end using a PC volume control. When $Z_L = Z_B$, any signal coming from the sound card speaker line through coils A and B of hybrid transformer 202 will establish voltages in coils H and G of hybrid transformer 204. These voltages will be equal and opposite in value (i.e. 180° out-of-phase), canceling each other out with resulting echo
 25 cancellation. Ideally, no signal originating from the sound card speaker line will be allowed to return back to its origin through the sound card microphone.

What has been shown and described reveals that the basis of hybrid transformer circuit 116 is the multiple windings or coils that make up the transformers. Voice signals are transferred between the same coils through electromagnetic coupling and, where couplings
 30 with opposite electromagnetic fields result, they will be canceled out. That is, by using the basic principle of the law of electromagnetism (equal but opposite-phase electromagnetic

fields), the voice echo generated by transmitting two different voice signal paths simultaneously can be canceled out. Therefore, two superposed alternating current (AC) signals can be transmitted by a single pair of wires and recovered individually without interfering with each other through the use of hybrid transformer circuit 116.

5 Referring specifically to FIG. 3, V_{TM} is the voltage from the telephone set microphone, V_{TS} is the voltage to the telephone set speaker, V_{RX} is the voltage from the sound card speaker output, and V_{TX} is the voltage to the sound card microphone input. When the speaker line is driven from the PC, then $V_{TS} = 0.55V_{RX}$ and $V_{TX} = 0$ as long as $Z_B = Z_L$. If Z_B were different from Z_L , then V_{TX} would be a non-zero value which results in a reflected
 10 echo signal back to the other end, which is undesirable. When the microphone line is driven, a voice signal from the telephone set is being injected into the microphone line with a voltage of $V_{TX} = 0.5V_{TM}$. Even though the voltage of the signal coming from the telephone set is reduced by one half, its level is still sufficient to be transmitted through the microphone line which can be amplified using a volume control on the PC on the other end.

15 Electromagnetic mutual coupling between hybrid transformers is reduced by improving the separation between the receiver and the transmitter in the four-wire circuit. This results in a reduction of the signal reflection. "Hybrid balance", which is terminology related to return loss, is the degree of electrical symmetry between two impedances connected to two conjugate sides of a hybrid circuit for minimizing cross-over signals
 20 between the transmit and receive sides. This is achieved by utilizing appropriate transformer turn ratios, appropriate core materials, as well as appropriate inductances. The symmetry of the signals transmitted and received is improved as impedance Z_B of the balance network is carefully equalized with the impedance Z_L from telephone set line for all transmitted frequencies of interest, as by using precise construction techniques in the transformer coils
 25 (i.e. precision in the ratio of winding turns during manufacture of the transformers).

The termination of impedance between the sound card and the cordless telephone is provided by coupling of the hybrid system on each side of the transmission/reception with an appropriate balance network. Leakage and inductances for the cores of the transformers and the capacitances between coils will result in reflection of signals at frequencies at which their
 30 impedances are significant. If these effects are ignored and the ratio of the coils is assumed

to be equal, the termination of impedances jointly with the appropriate balance network will achieve echo cancellation and a clear and comfortable voice conversation over the Internet.

The first embodiment described in relation to FIGs. 2-3 utilizes hybrid transformers having relatively large cores, such as SPT-180-type transformers previously described. However, a second embodiment uses transformers with relatively smaller cores, such as SPT-060 type transformers which are also available from Prem Magnetics, Inc. or Model TL028 transformers from Mouser Electronics. This second embodiment is shown and described in relation to FIG. 4.

The hybrid transformer circuit 116 of FIG. 4 includes a hybrid transformer 402, a hybrid transformer 404, and an impedance matching circuit 406. Hybrid transformer 402 includes a coil A which is inductively coupled to both coils B and C formed along the same core. Similarly, hybrid transformer 404 includes a coil F which is inductively coupled to both coils D and E formed along the same core. Terminals 1 and 2 of coil A of hybrid transformer 402 are coupled to speaker plug 122 (FIG. 1), whereas terminals 1 and 2 of coil F of hybrid transformer 404 are coupled to microphone plug 124 (FIG. 1). A terminal 5 of coil C of hybrid transformer 402 and a terminal 6 of coil E of hybrid transformer 404 are coupled to two-line telephone jack 118 (FIG. 1). A terminal 6 of coil C of hybrid transformer 402 is coupled to a terminal 5 of coil E of hybrid transformer 404. A terminal 4 of coil B of hybrid transformer 402 is coupled to a terminal 4 of coil D of hybrid transformer 404, making coils B and D reverse-coupled. A terminal 3 of coil B of hybrid transformer 402 and a terminal 3 of coil D of hybrid transformer 404 are coupled to impedance matching circuit 406 which has an impedance Z_B . In this embodiment, impedance matching circuit 406 includes a resistor 410 and an optional capacitor 408 (optional) coupled in a parallel configuration as shown. The impedance Z_B , capacitor 408, and resistor 410 may be the same or similar components as described in relation to FIGs. 2-3.

Although the physical dimensions of the two transformers of FIG. 4 are different, their behavior and performance in this application are very similar. And although the number of coils in the second configuration of FIG. 4 differs by two (one in transmitting part Tx and the other in receiving part Rx), the explanation described above in relation to FIGs. 2-3 is applicable here as well. For the purposes of mass production of the interface adapter, it has been considered more suitable to use a configuration with SPT-060 hybrid transformers

(Prem Magnetics) or TL028 hybrid transformers (Mouser Electronics) due to the great economic savings it represents for manufacturing relative to the use of the SPT-180 transformers. In this embodiment, the ratio N for both transformers 402 and 404 is between 1:1.60 (+/- 2%) for the SPT-060 type.

5 FIG. 5 is a schematic diagram of the hybrid transformer circuit 116 of FIG. 4 which further includes a voltage biasing interface 502. Although shown with respect to FIG. 5, voltage biasing circuit 502 may be utilized in hybrid transformer circuit 116 of FIGs. 2 and 3 as well. Voltage biasing interface 502 is utilized for biasing circuits of some cordless telephone sets which require it; however not all cordless telephone system will require such
10 biasing. In general, voltage biasing interface 502 is adapted for coupling to a voltage source which provides a bias voltage to the cordless telephone system through hybrid transformer circuit 116. More particularly, voltage biasing interface 502 utilizes a USB plug 504 and cable which is coupled to one of the lines of the telephone interface (RJ-11 female jack) of the interface adapter. As particularly shown in FIG. 5, USB plug 504 is coupled through
15 terminal 5 of coil C of hybrid transformer 402. USB is a high-performance networking standard which is based on a serial bus architecture. Most new computers and associated peripheral devices like printers and scanners support USB.

 USB plug 504 is adapted to be connected to a corresponding USB jack on the computer/PC. The voltage bias obtained from the USB is about 5 volts (+5 volts line and
20 ground line), which provides a sufficient voltage bias for most (if not all) of such cordless telephone sets. The USB cable extends from housing 120 (FIG. 1) in a manner similar to cables 126 and 128 (FIG. 1). Preferably, as described above, the USB cable and cables 126 and 128 (FIG. 1) are integrally combined (bundled) along a single cable strand and split at the end into separate connectors. Note that only positive and negative terminals of USB plug
25 504 are coupled to hybrid transformer circuit 116; all other lines (e.g. data lines) of USB plug 504 are not utilized. A capacitor 508 of the electrolytic type is coupled between the telephone set line and terminal 5 (coil C) of hybrid transformer 402 as shown in FIG. 5; the positive and negative terminals from USB plug 504 are coupled across this capacitor 508. Also, a current-limiting resistor 506 is coupled along the positive terminal as shown in FIG.
30 5. Preferably, the resistance of resistor 506 is about 62 Ω (rated at ¼ Watt) and the capacitance of capacitor 508 is about 470 μ F (rated at 6.3 volts). It is important to note that,

even though voltage biasing interface 502 is utilized in this embodiment, hybrid transformer circuit 116 of FIG. 5 still consists entirely of passive components.

Turning now to FIG. 6, an illustration of hybrid transformers 202 and 204 of the hybrid transformer circuit in housing unit 120 is shown. Hybrid transformers 202 and 204 are mounted and coupled on a printed circuit board (PCB) 602 which is contained within housing unit 120. Hybrid transformers 202 and 204 must be positioned relatively close together on PCB 602 to ensure that the size of the interface adapter is small. However, hybrid transformer 202 has electromagnetic waves 604 that radiate primarily out of one side thereof, and hybrid transformer 204 similarly has electromagnetic waves 606 that radiate primarily out of another side thereof. Positioning hybrid transformers 202 and 204 close to one another would tend to cause interference, which would provide an impedance mismatch and increase voice echo.

In the present application, however, hybrid transformers 202 and 204 are oriented so as to be perpendicular to one another (i.e. rotated 90° out-of-phase with each other) to reduce or minimize mutual electromagnetic coupling. Thus, hybrid transformers 202 and 204 may still be positioned relatively closely on PCB 602 to conserve space within housing unit 120 without electromagnetically interfering with one another. In this embodiment, hybrid transformers 202 and 204 each have dimensions of about ½" by ½" (1.27 cm), are spaced apart by about 1 inch (2.54 cm), and the dimension of PCB 602 is about 1.1 by 2.3 inches (2.8 by 5.8 cm). As apparent, the interface adapter can be made relatively smaller in size without adversely affecting the performance of the adapter. Although being rotated 90° out-of-phase is preferred for good results, any suitable rotation between 45° – 135° may be utilized to reduce undesirable electromagnetic coupling. Thus, hybrid transformer 202 and 204 are physically oriented within housing unit 120 so as to reduce electromagnetic interference with one another for conserving space within the housing unit 120.

What has been described herein is a cordless telephone-to-sound card interface adapter that will allow Internet users to establish PC-to-telephone and PC-to-PC voice conversations over the Internet by means of instant messaging using their PC and any standard cordless telephone, freeing them from sitting idly in front of their PC and offering mobility while they talk anywhere in their homes. Most preferably, the interface adapter utilizes 100% passive components for its operation, which makes it more affordable and

leads to fewer risks of electrical discharges. The interface adapter consists of a hybrid transformer circuit for matching the impedance between a PC sound card (600 Ω impedance from the speaker and 10K Ω impedance from the microphone) and the cordless telephone (nominal impedance of 450 Ω), thereby minimizing and/or canceling out voice echo during
5 telephone conversations over the Internet. Advantageously, the interface adapter provides a low-cost yet effective means to provide mobility for an end user during voice communications over the Internet. With convenient cordless telephone mobility, this adapter facilitates significant savings in long distance telephone services.

The above-described embodiments of the present application are examples only.
10 Those of skill in the art may effect alterations, modifications and variations to the particular embodiments without departing from the scope of the application. For example, although the use of a cordless telephone system is a primary focus of the present application, the interface adapter of the present invention may be utilized with any standard telephone set. Finally, the invention described herein in the recited claims intend to cover and embrace all suitable
15 changes in technology.

What Is Claimed Is: